

## 4 Technical specifications

### 4.1 Range

In practice the range of this type of specific resistivity meter is restricted for the following reasons.

(i) The maximum current supply cannot be much higher than 1 A without requiring transformers too large for this kind of instrument.

(ii) The input impedance at the preamplifier is in the region of 10 M $\Omega$  due to the capacitive connections at the measuring frequency.

(iii) 10 mV is the smallest detectable voltage signal able to be handled by the AC-logarithmic circuits.

Conditions (i) and (iii) indicate that the lowest measurable resistances of the sample between the potential contacts could be of the order of 10 m $\Omega$ . Condition (ii) implies that the highest measurable resistances could be of the order of 10 M $\Omega$ . Thus with these simple assumptions the range of the device would be nine decades. With special arrangements it is possible to extend the range of the meter by a few decades (Andersson *et al* 1975).

Because the difference between the signal levels of  $u_0$  and  $u$  cannot be much higher than three decades in practice, one reference resistance is not enough to cover the whole range. For simplicity it was decided that two reference resistances with a manual scale-changing selector would be used.

### 4.2 Calibration

Calibration of the instrument is performed easily and accounts for the  $k$  factor outlined in equation (1) and table 1. The measuring circuit is made compatible with all sample holders by choosing  $R_0$  so that  $k$  is equal for all sample holders. Calibration can also be done experimentally by replacing the sample with suitably selected resistor chains.

### 4.3 Accuracy

Because of the purpose of the measurement the meter was not designed to be a high-accuracy instrument. Through the nine decades the accuracy is better than  $\pm\frac{1}{2}$  decade. The main reasons for this inaccuracy are the non-ideal logarithmic amplifier and the various stray currents, especially capacitive ones, in the sample holder. However, it is possible to increase the accuracy and extend the range of the meter easily by eliminating stray currents and by enlarging the signal levels.

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## System for the measurement of fast response times under fast electromagnetic irradiation

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**Abstract** We describe a system that has been used successfully for the measurement of relaxation times of superconducting tunnel junctions under light irradiation. The system allows the measurement of 2  $\mu$ V signals of less than 10 ns rise time, with a signal-to-noise ratio of better than 4. We also describe, for the first time, the behaviour of a commercial GaAs solid state laser at cryogenic temperatures.

### 1 Introduction

Recently there has been considerable interest in the time-dependent properties of solids, particularly in the field of non-equilibrium superconductivity. The basic parameters being studied are the various relaxation times involved. We describe a system that allowed the measurement (Schuller and Gray 1976, 1977) of 2  $\mu$ V signals with a signal-to-noise ratio of 4 and rise times of less than 10 ns, generated by a superconducting tunnel junction that was irradiated with a fast laser pulse.

We also describe, we believe for the first time, the behaviour of a commercial GaAs solid state laser (904 nm) at cryogenic temperatures. In any laboratory that has cryogenic facilities, the use of a GaAs laser can save a considerable amount of money where a high-power pulsed laser is required.

### 2 Apparatus

The irradiation system consisted of a GaAs injection laser (RCA, SG2012) pulsed by a fast pulser (HP214A) of about 10 ns rise time. The detection system used fast amplifiers and a signal averager. The tunnel junction and the laser were encased in separate superconducting niobium shields to avoid direct electromagnetic pick-up (seen as a ringing at the beginning and end of the pulse), and they were located either in the same <sup>4</sup>He Dewar (Schuller and Gray 1976) or in two separate Dewars (Schuller and Gray 1977). To improve the screening still further, 50  $\Omega$  solid coaxial lines (UT-141SS, available from Blake Assoc., Medford, Massachusetts) were used for the electrical pulses to the laser and for the biasing and detecting channels. The 115 V supplies for the amplifiers, oscilloscopes, etc were floated to eliminate all possible ground loops. The total pick-up at the junction was less than 15  $\mu$ V peak-to-peak. A block diagram of the experimental set-up is shown in figure 1. The junction was optically connected to the laser by a fibre optic bundle that had a loss of less than  $\frac{1}{2}$  dB. The junction was biased with a battery-operated power supply, and the high-frequency pulses were blocked from this supply with a low-pass RC filter.

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### Apparatus and techniques

In general the performance improves at lower temperatures, particularly below the  $\lambda$  point of helium. This is probably due to the fact that at helium temperatures the heat is conducted away from the laser more efficiently than at higher temperatures. A comparison of the performance of one of the lasers at different temperatures is given in table 1. A systematic study of the operation of this laser at helium temperatures can provide more information (Ciftan and Debye 1965, Keyes 1970, Chakravarti and Parui 1971, Eliseev 1973, Litvinov *et al* 1974) in problems relating to the operation and degradation due to thermal processes in semiconductor lasers.

**Table 1** Comparison of laser parameters at different temperatures.

	77 K	4.2 K	1.3 K
Maximum pulse width ( $\mu$ s)	0.7	30	150
Lasing threshold (mA)	600	300	200
Forward voltage (V)	2.0	1.5	0.7
Maximum duty factor (%)	5	12	25

One of us (IS) is presently studying the behaviour of superconductors under microwave irradiation. The detection system is essentially the same. The microwaves are pulsed using an SPST pin diode switch (Narda Microwave Corp., Plainview, Long Island, New York 11803) with a bandwidth of 0.2–18 GHz. The results will be published elsewhere (Schuller and Dahlberg 1979).

In summary, we describe a system capable of detecting microvolt signals with nanosecond rise time using signal-averaging techniques.

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